



TRW



Continuous Phase Modulation For The Indoor Wireless Channel

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Motivation

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- Indoor communication is increasingly important (computer networks, remote control & sensing).
- CPM yields increased spectral efficiency; increases system capacity.
- Understand CPM performance in the severe channel of multipath fading and cochannel interference.



Continuous Phase Modulation

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■ Transmitted signal

$$s(t) = \sqrt{\frac{2E}{T}} \cos[\omega_c t + \phi(t, \alpha)]$$

■ Information phase

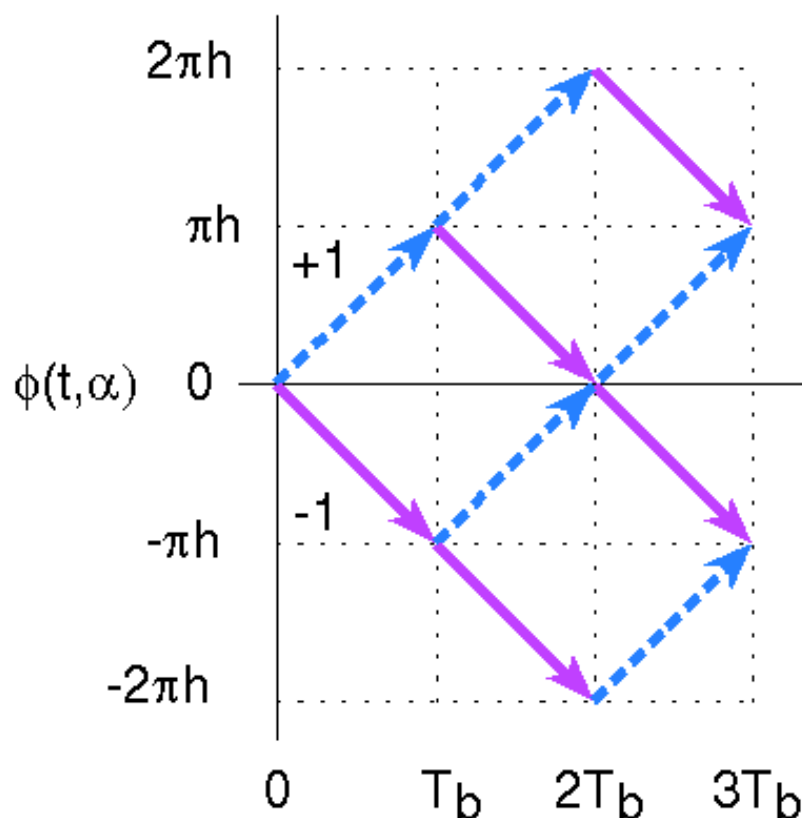
$$\phi(t, \alpha) = 2\pi h \sum_{i=-\infty}^{\infty} \alpha_i q(t - iT_b)$$

■ Phase function - 1REC

$$q(t) = \left\{ \begin{array}{ll} 0, & t < 0 \\ \frac{t}{2T_b}, & 0 \leq t \leq T_b \\ \frac{1}{2}, & t \geq T_b \end{array} \right\}$$



■ Restrict modulation index

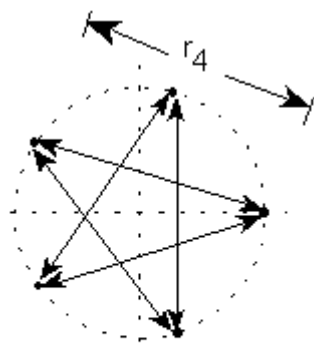


$$h = \frac{2n}{m}, \quad n, m = 1, 2, 3, \dots$$

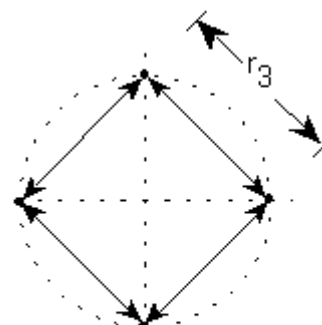
$$h = \frac{2}{3}, \frac{1}{2}, \frac{4}{5}, 1, \dots$$

The number of trellis states is m .

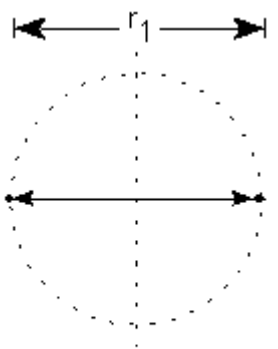
CPM Constellations



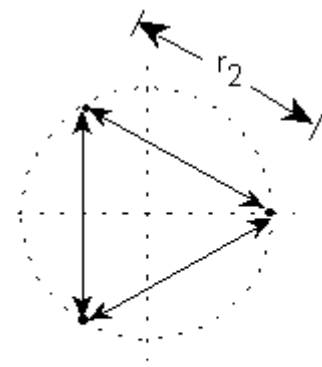
$h=4/5$



$h=1/2$



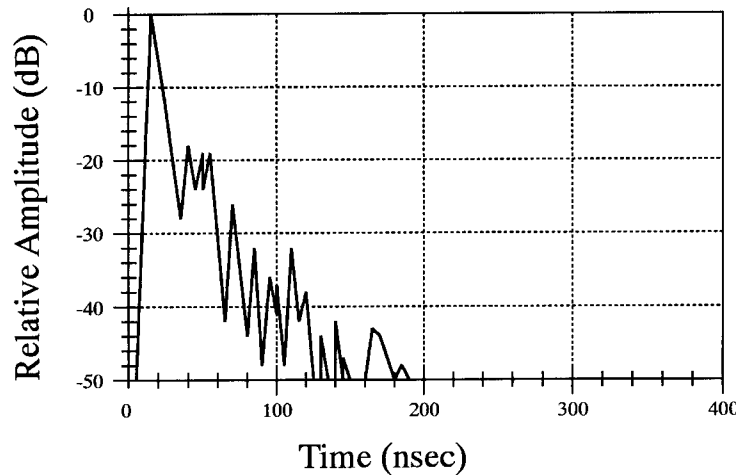
$h=1$



$h=2/3$



■ Channel impulse response



$$h(t) = Ae^{j\mu} \delta(t - \zeta) + \sum_{l=1}^L \bar{\beta}_l e^{j\bar{\gamma}_l} \delta(t - \bar{\tau}_l)$$

$$E[\bar{\beta}_l^2] = 2\rho_l$$

Homayoun Hashemi, "The Indoor Radio Propagation Channel", *IEEE Proceedings*, Vol. 51, No. 7, July 1993.

■ Rician factor 6-12 dB

$$H_l = \frac{A^2}{2\rho_l}$$

- Maximum Likelihood Sequence Estimation maximizes the likelihood function up to the n th symbol

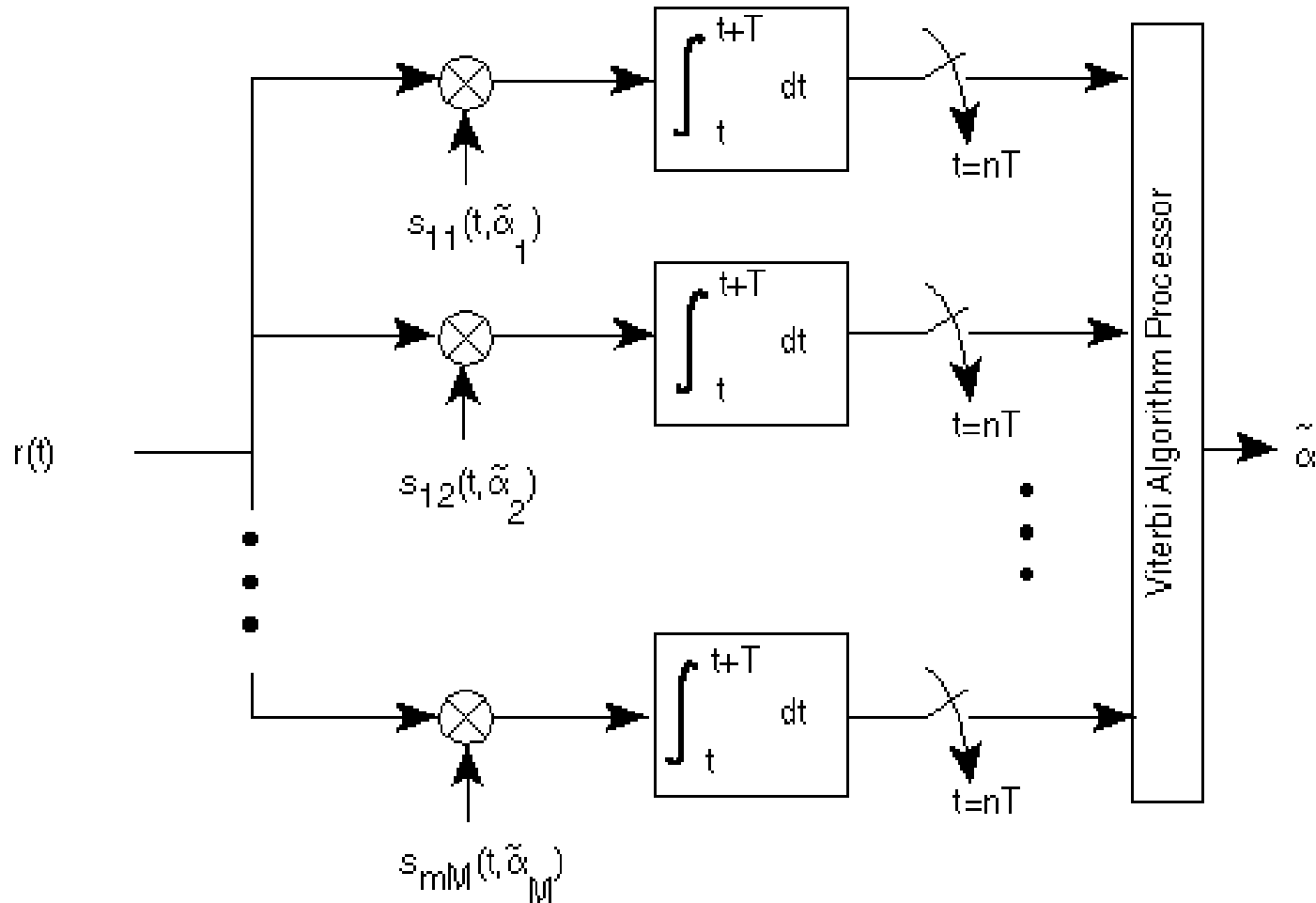
$$J_n(\vec{\alpha}) = \int_{-\infty}^{(n+1)T} r(t)s(t, \vec{\alpha})dt,$$

or equivalently written

$$J_n(\vec{\alpha}) = J_{n-1}(\vec{\alpha}) + Z_n(\vec{\alpha}),$$

where

$$Z_n(\vec{\alpha}) = \int_{nT}^{(n+1)T} r(t)s(t, \vec{\alpha})dt$$





Received Signal

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$$\begin{aligned} r(t) = & A\sqrt{2P} \cos[\omega_c t + \phi(t, \vec{\alpha})] \\ & + \sum_{k=1}^K A\sqrt{2P_k} \cos[\omega_c (t - \bar{\tau}_k) + \phi(t - \bar{\tau}_k, \vec{\alpha}_k) + \bar{\theta}_k] \\ & + n(t) \end{aligned}$$

Where $n(t)$ is conditional Gaussian, and is composed of AWGN, desired signal multipath and interference multipath random components.

- Based on *Error Events* diverging from the correct trellis path and merging after some symbol periods

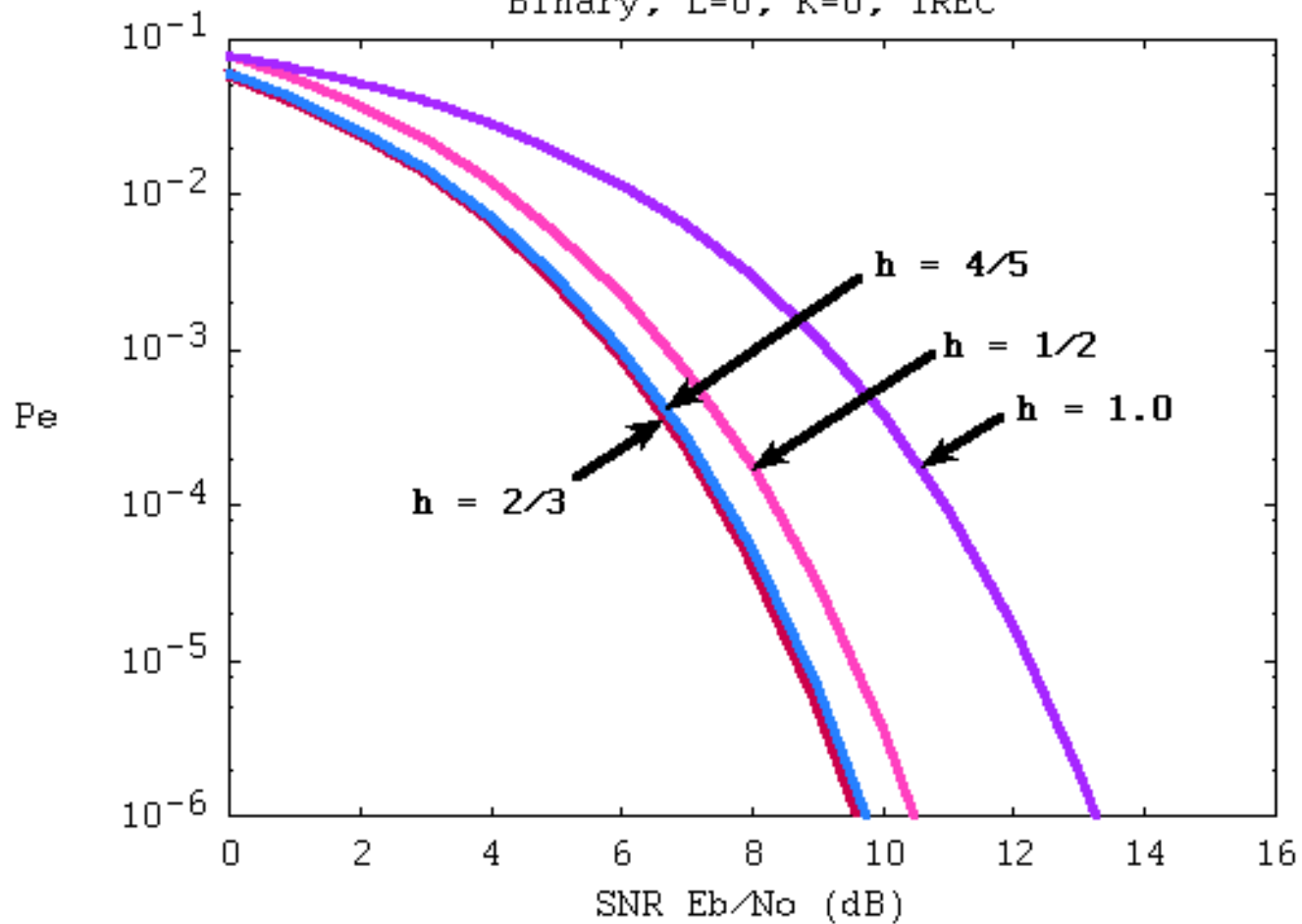
$$\Pr[E] = \sum_{\alpha_m} \sum_{\tilde{\alpha}_m \neq \alpha_m} w(\alpha_m, \tilde{\alpha}_m) \prod_{i=0}^{m-1} \frac{M - |\epsilon_i|}{M} \frac{1}{2^{K(m+1)}} \sum_{k=1}^{2^{(m+1)}} \frac{1}{T^K} \frac{1}{T^L} \frac{1}{T^{KL}} \frac{1}{(2\pi)^K}$$

$$\int_0^T \int_0^T \int_0^T \int_0^{2\pi} Q \left(\frac{d_s^2(\alpha_m - \tilde{\alpha}_m) + x_I(\alpha_m, \tilde{\alpha}_m, \alpha_k, \bar{\theta}_K, \bar{\tau}_K)}{\sqrt{\sigma_{N_m}^2(\alpha_m, \tilde{\alpha}_m, \alpha_k, \bar{\theta}_K, \bar{\tau}_K, \bar{\tau}_L, \bar{\tau}_{KL})}} \right) d\bar{\theta}_K d\bar{\tau}_K d\bar{\tau}_{KL} d\bar{\tau}_L$$



No Interference or Fading **TRW**

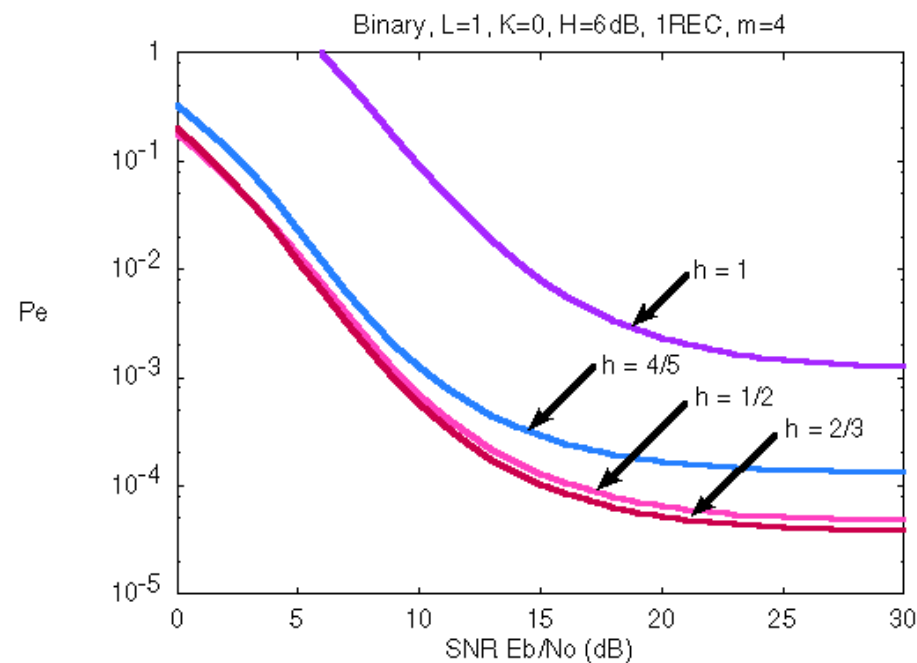
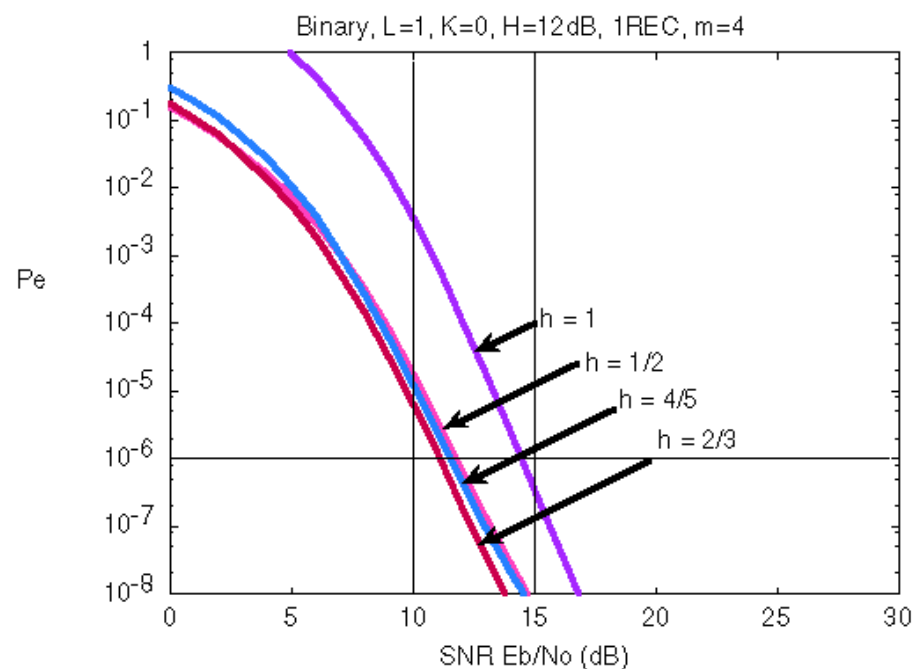
Binary, $L=0$, $K=0$, 1REC





Single Multipath

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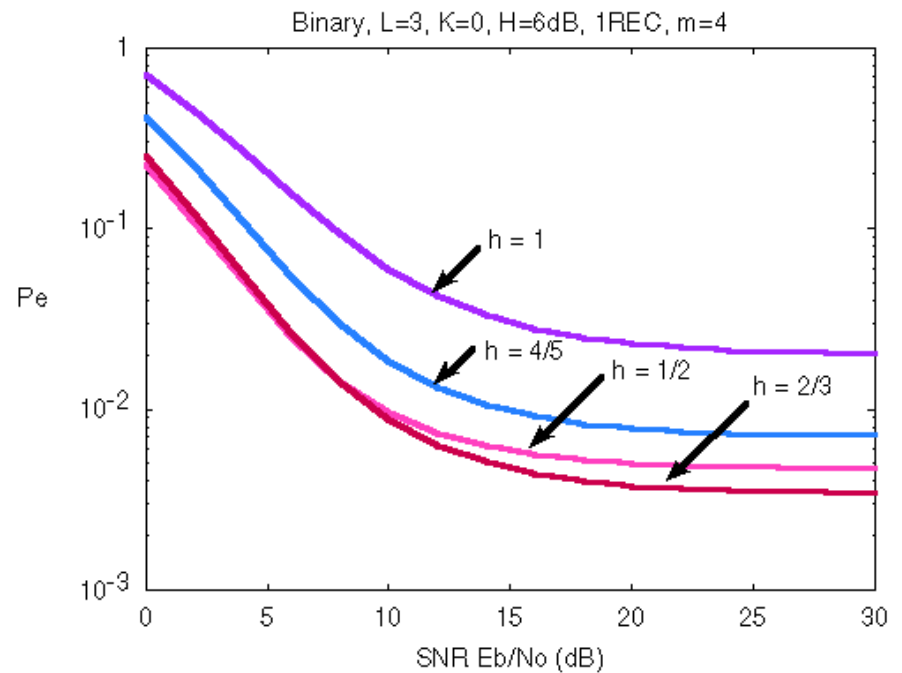
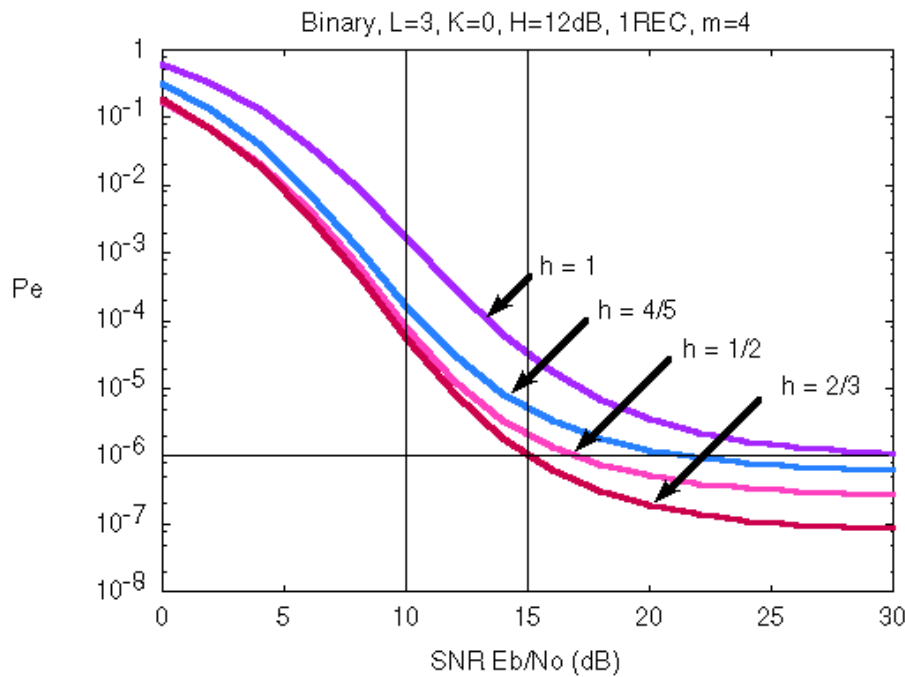


$$H = \frac{A^2}{2\rho_1}$$



Three Constant Power Multipaths

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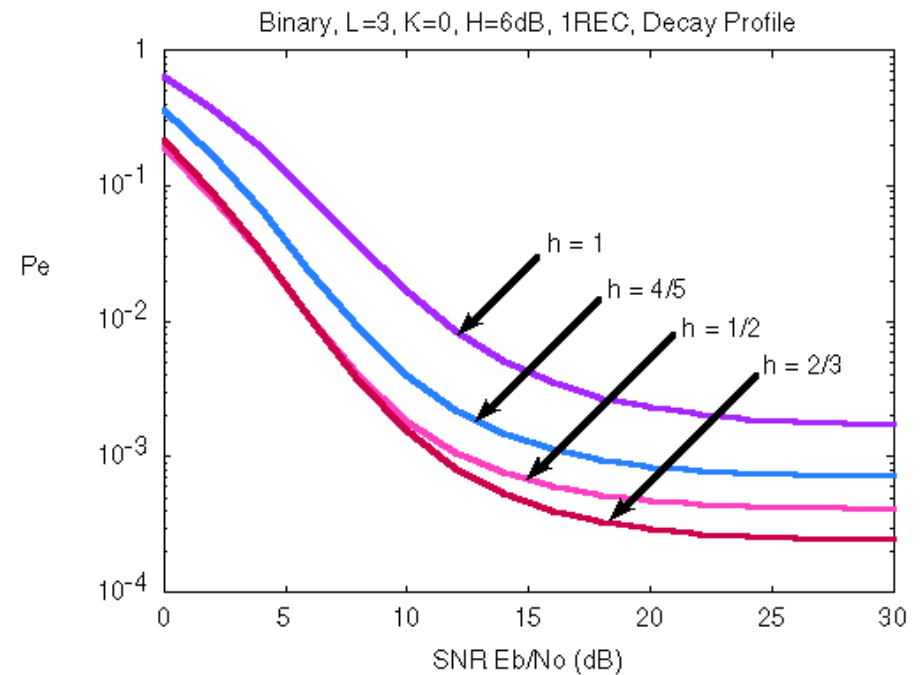
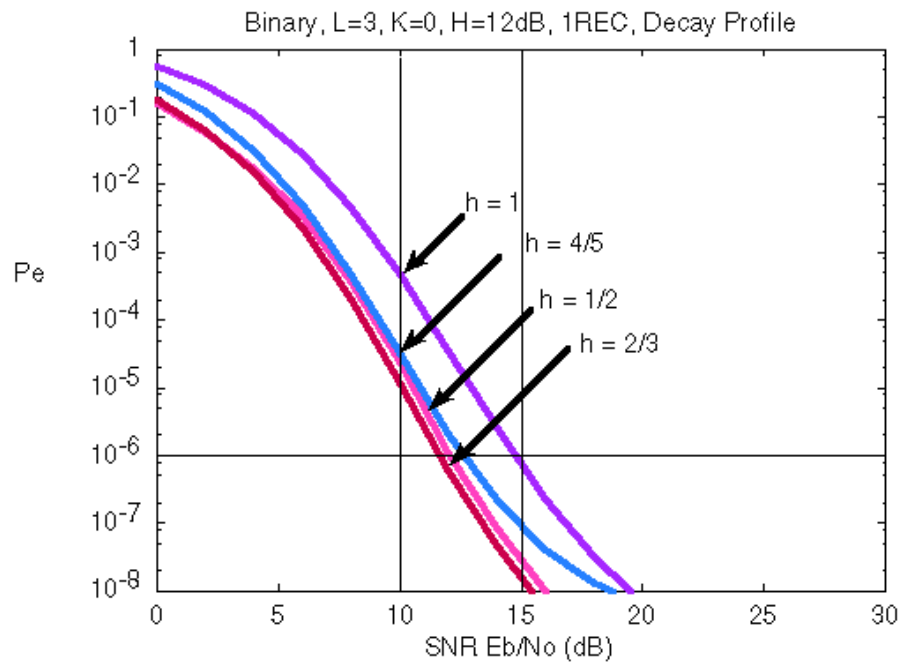
$$H = \frac{A^2}{2\rho_l}, \quad l = 1, 2, 3$$

$$\rho_1 = \rho_2 = \rho_3$$



Three Decaying Multipaths

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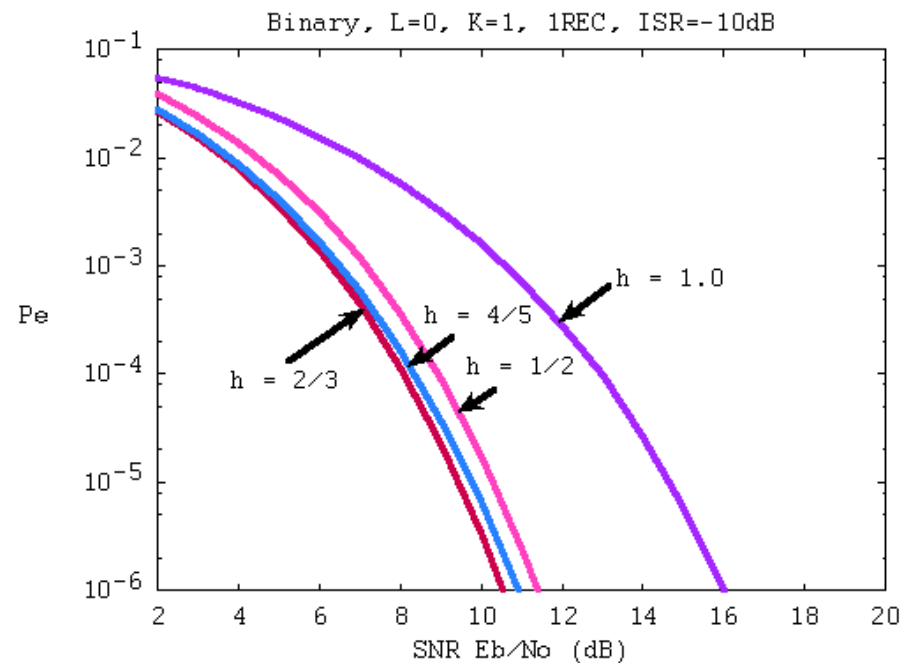
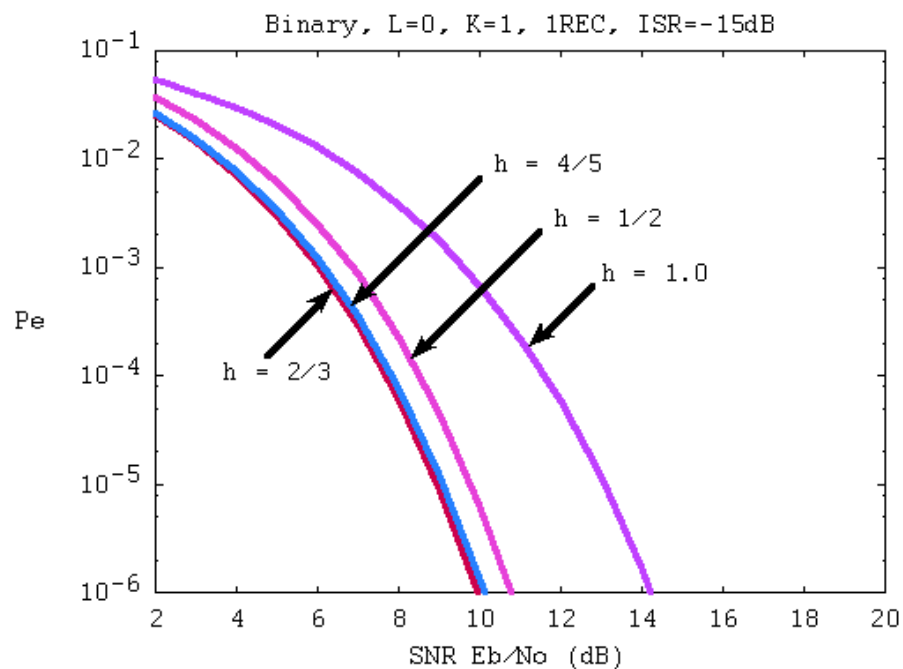


$$H = \frac{A^2}{2\rho_1}, \quad \rho_1 = 2\rho_2 = 4\rho_3$$



Single Interferer

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$$ISR = \frac{P_1}{P}$$

3-Feb-98

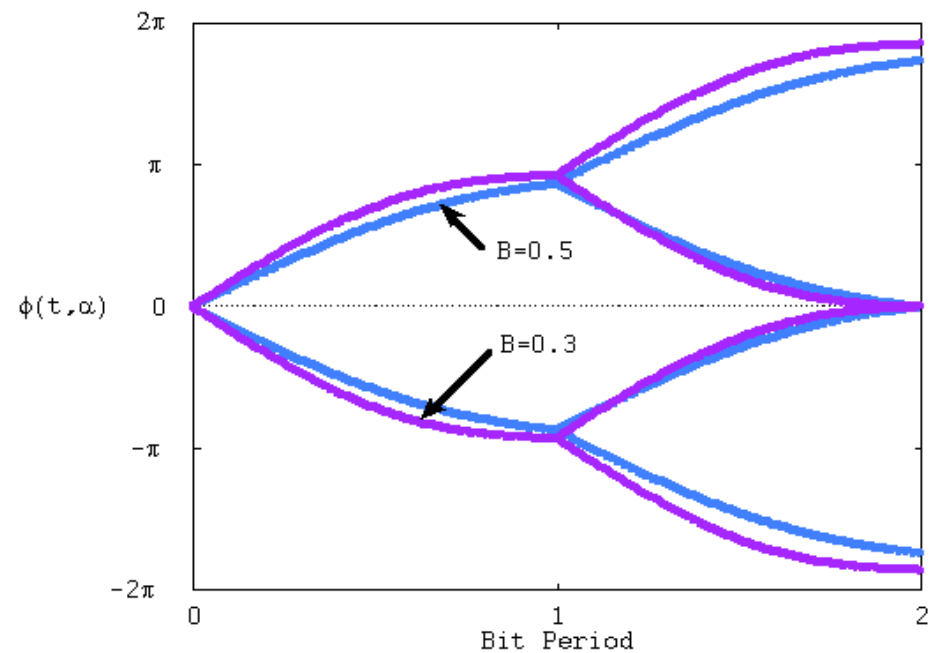
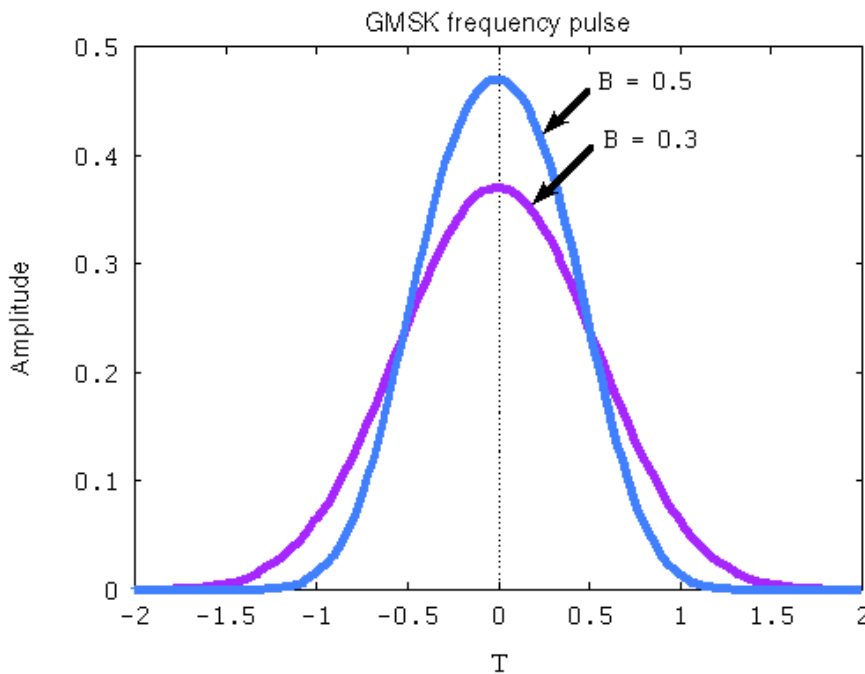
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Gaussian Filtered MSK **TRW**



- Gaussian frequency pulses, resulting in arced trellis transitions.





GMSK

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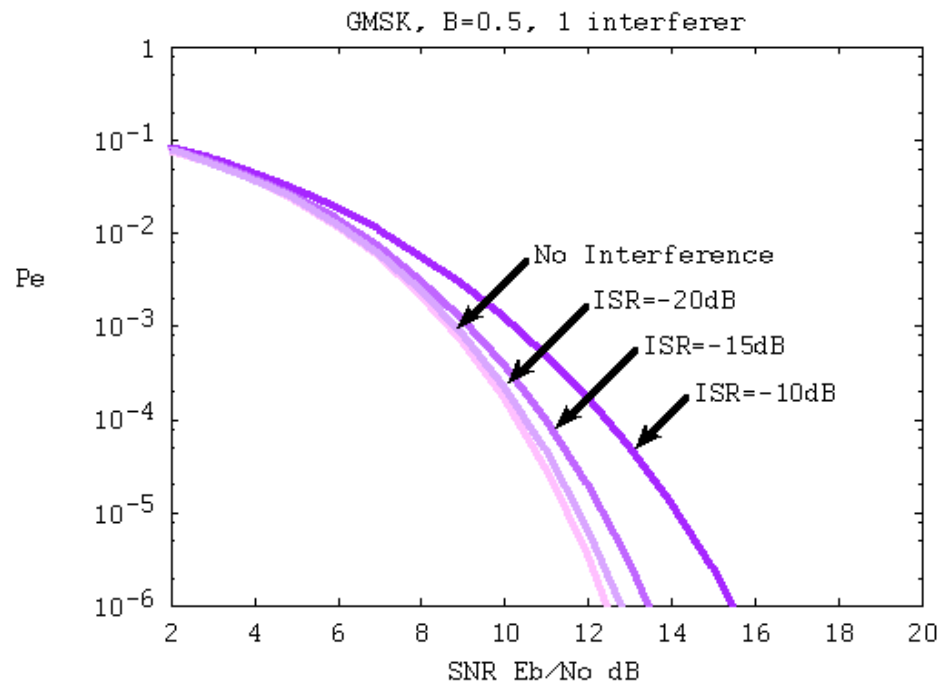
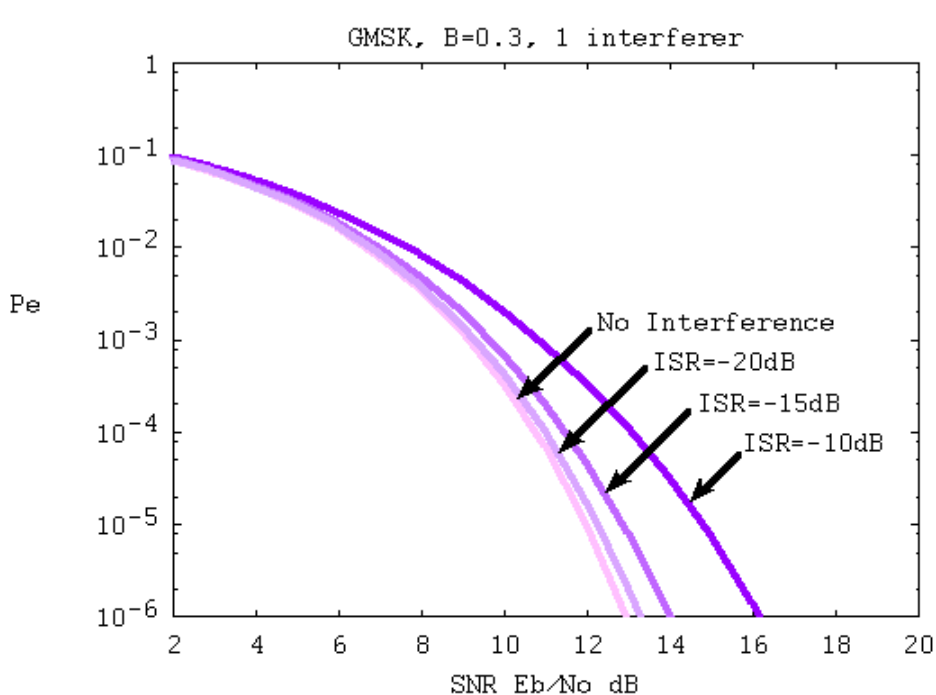


- GMSK standards established for
 - ◆ Global System for Mobile (GSM) Communications [B=0.3]
 - ◆ Digital European Cordless Telephone (DECT) [B=0.5]
 - ◆ Digital Communication System (DCS) [B=0.3]
- Compare our error performance results with these systems.



GMSK Performance

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$$ISR = \frac{P_1}{P}$$



Monte-Carlo Integration **TRW**

- If $\{x_1, x_2, \dots, x_N\}$ are independent random variables with density $p(x)$ then

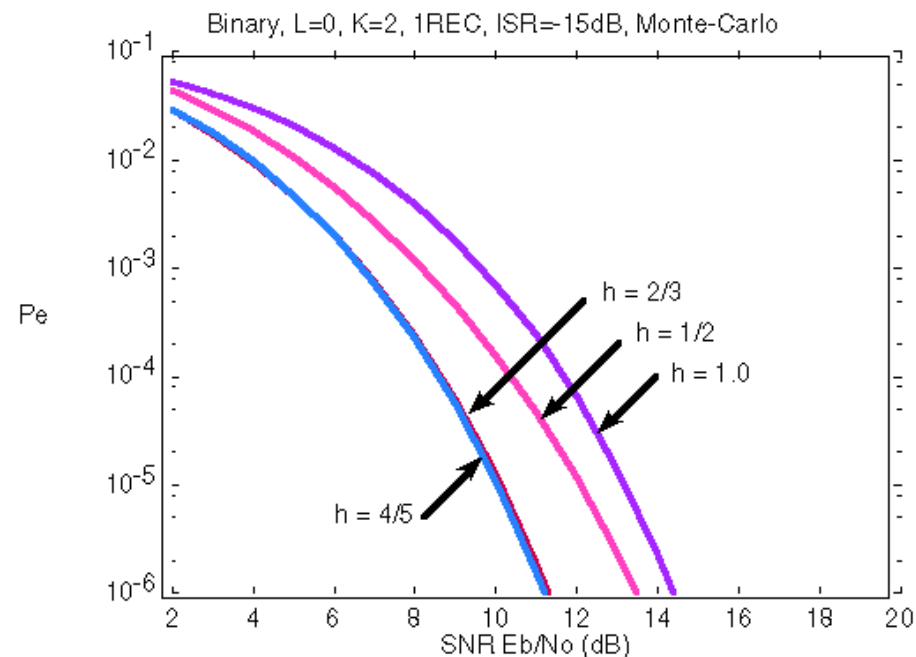
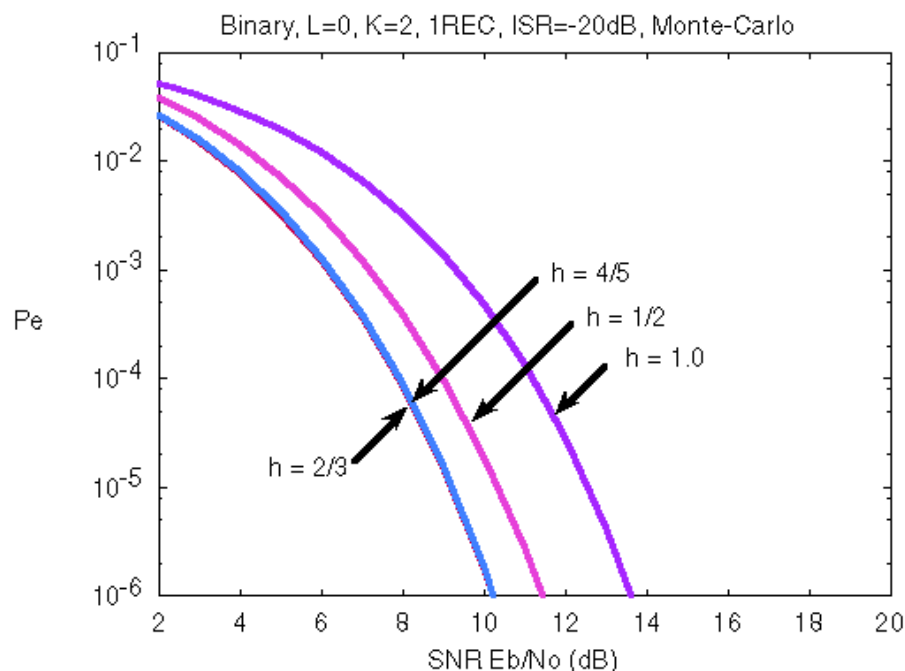
$$\int f(x) p(x) dx = E[f(x)] \approx \frac{1}{N} \sum_{n=1}^N f(x_n)$$

- Extension to multidimensional integrals results in rapid computation.
- The accuracy is independent of the integration dimension.



Two Interferers

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$$ISR = \frac{P_k}{P}, k = 1, 2 \quad P_1 = P_2$$

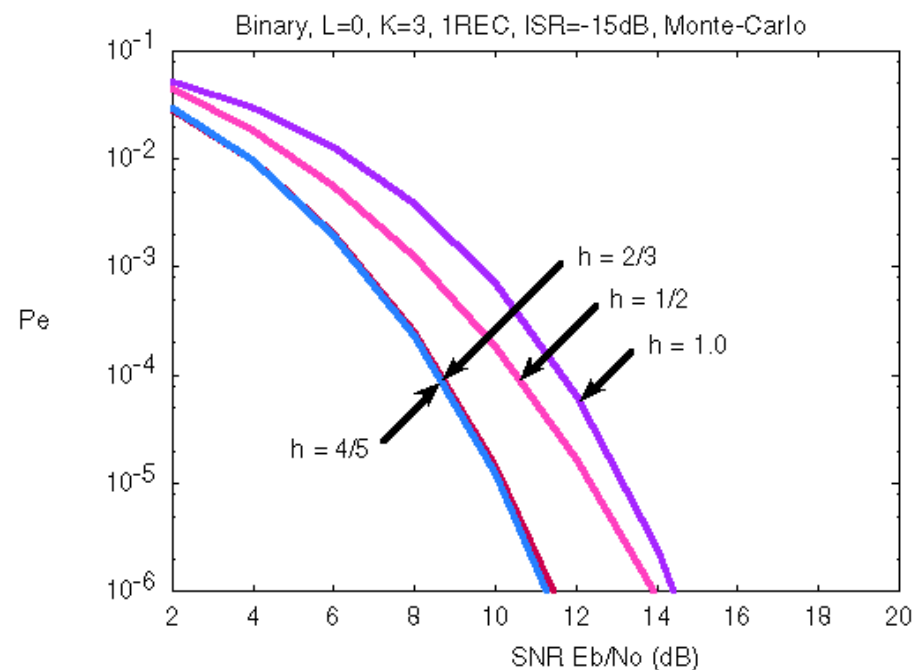
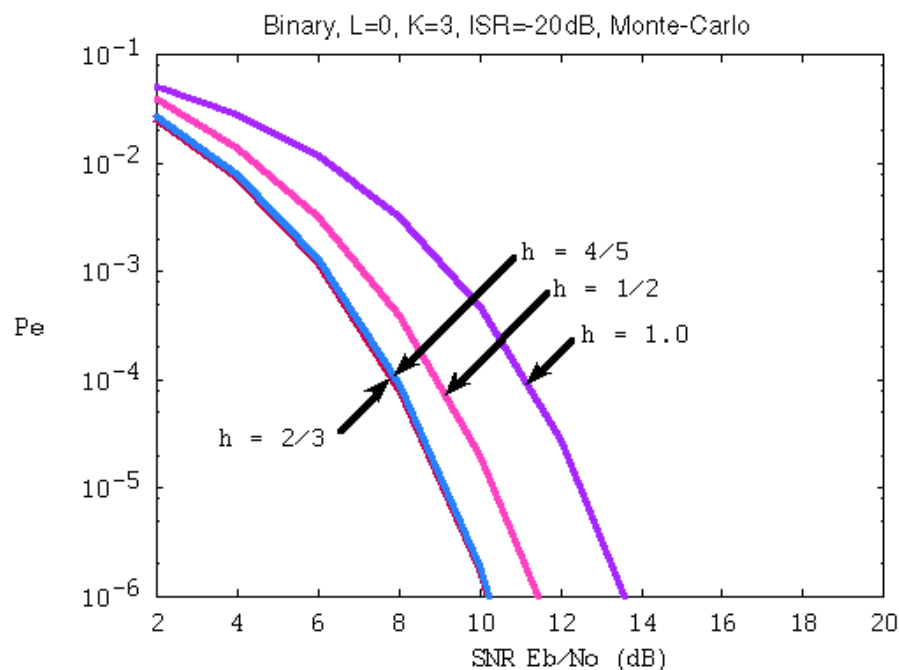
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Three Interferers

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$$ISR = \frac{P_k}{P}, k = 1, 2, 3 \quad P_1 = P_2 = P_3$$



Conclusions

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- 
- A horizontal bar composed of several rectangular segments in various shades of blue, ranging from dark to light, with some segments having a gradient effect.
- Complete error performance expression obtained.
 - Reduced computation by using the Monte-Carlo integration method.
 - “Best” modulation index is $h=2/3$ under a variety of conditions, and has a simple three state MLSE decoder.